

## Overview

The LA9702W is an RF signal-processing and servo error signal generation IC for DVD and CD playback. A DVD player can be implemented by combining this IC with a DVD DSP product that includes a digital servo DSP.

## Functions and Features

- RF signal generation (with built-in gain adjustment VGA circuit)
- RF peak detection and generation
- RF bottom detection and generation (with time constant switching)
- Built-in RF equalizer amplifiers (two systems)
- FE amplifier (with built-in balance adjustment VCA and offset cancellation pin)
- Three-beam TE amplifier (with built-in balance adjustment VCA and offset cancellation pin)
- Reflection amplifier
- DPD circuit
- Tracking hold circuit
- Push-pull TE amplifier (with built-in balance adjustment VCA)
- Built-in wobble detection bandpass filter
- APC circuits (two systems)
- Defect detection circuit


## Specifications

Maximum Ratings at $\mathbf{T a}=25^{\circ} \mathrm{C}$

## Package Dimensions

unit: mm
3220-SQFP80


| Parameter | Symbol | Conditions | Ratings |  |
| :--- | :---: | :---: | :---: | :---: |
| Maximum supply voltage | $\mathrm{V}_{\text {CC }}$ max |  | 6.0 | V |
| Allowable power dissipation | Pd max | Pdmax for $\leq 70^{\circ} \mathrm{C}($ when mounted on a PCB) |  |  |
| Operating temperature | Topr |  | 500 | mW |
| Storage temperature | Tstg |  | -25 to +70 | ${ }^{\circ} \mathrm{C}$ |
| ${ }^{\circ} \mathrm{C}$ |  |  |  |  |

Note: These specifications are subject to change without notice.

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Operating Conditions at $\mathbf{T a}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Recommended supply voltage | Vcc |  | 5.0 | V |
| Operating supply voltage | Vccop |  | 4.5 to 5.5 | V |

Electrical Characteristics/Operating Characteristics
at $\mathbf{T a}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}($ pins 12,55 , and 66$)=5 \mathrm{~V}$, ground (pins 17, 49, and 74) $=0 \mathrm{~V}$

| Parameter | Symbol | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | min | typ | max |  |
| Current drain | ICC | No signal | 40 | 58 | 75 | mA |
| Reference voltage 1 | PREF | Pin 79, load current: $\pm 2 \mathrm{~mA}$ | 2.3 | 2.5 | 2.7 | V |
| Reference voltage 2 | SREF | Pin 27, load current: $\pm 2 \mathrm{~mA}$ | 2.3 | 2.5 | 2.7 | V |
| RF gain 1 | RFG1 | Input to pins 1 and 2, pin $21=$ SREF -0.75 V Pin $61=$ pin $63=$ pin $76=5 \mathrm{~V}$, pins 40 and 41 | 0.25 | 3.25 | 6.25 | dB |
| RF gain 2 | RFG2 | Input to pins 1 and 2, pin $21=\mathrm{REF}+0.75 \mathrm{~V}$ Pin $61=$ pin $63=$ pin $76=5 \mathrm{~V}$, pins 40 and 41 | 18 | 21 | 24 | dB |
| PH | PH | $\begin{aligned} & \text { Pin } 7=\operatorname{pin} 8=\operatorname{pin} 9=400 \mathrm{mV} \text { pp, pin } 21=\text { SREF }-0.75 \mathrm{~V} \\ & \text { Pin } 61=5 \mathrm{~V} \text {, pin } 63=\operatorname{pin} 76=0 \mathrm{~V} \text {, pin } 44 \end{aligned}$ | SREF + 0.4 | SREF + 0.65 | SREF + 0.9 | V |
| BH | BH | $\begin{array}{\|l} \text { Pin } 7=\operatorname{pin} 8=\operatorname{pin} 9=400 \mathrm{mV} \text { pp, pin } 21=\text { SREF }-0.75 \mathrm{~V} \\ \text { Pin } 57=\operatorname{pin} 59=\operatorname{pin} 61=5 \mathrm{~V} \text {, pin } 63=\operatorname{pin} 76=0 \mathrm{~V} \text {, pin } 42 \\ \hline \end{array}$ | SREF - 0.08 | SREF - 0.28 | SREF - 0.48 | V |
| RREC1 | RREC1 | Input to pins $3,4,5$, and 6 , pin $18=$ SREF -0.75 V , pin $76=5 \mathrm{~V}$, Pin 28 | 4.2 | 9.2 | 14.2 | dB |
| RREC2 | RREC2 | Input to pins $3,4,5$, and 6 , pin $18=$ SREF -0.75 V , pin $76=5 \mathrm{~V}$, Pin 28 | 20 | 23 | 26 | dB |
| RRECOST | ROST | Pin $3=\operatorname{pin} 4=\operatorname{pin} 5=\operatorname{pin} 6=$ PREF, $\operatorname{pin} 18=$ SREF Pin $76=5 \mathrm{~V}$, pin 28 | SREF - 0.3 | SREF | SREF + 0.3 | V |
| FEGAIN1 | FEG1 | Input to pins $3,4,5$, and 6 , pin $18=$ SREF -0.75 V , pin $76=5 \mathrm{~V}$, Pin 29 | 13.9 | 17.9 | 21.9 | dB |
| FEGAIN2 | FEG2 | Input to pins $3,4,5$, and 6 , pin $18=$ SREF +0.75 V , pin $76=5 \mathrm{~V}$, Pin 29 | 24 | 27 | 30 | dB |
| FEOST | FOST | Pin $3=\operatorname{pin} 4=\operatorname{pin} 5=\operatorname{pin} 6=$ PREF, pin $18=$ SREF Pin $76=5 \mathrm{~V}$, pin 29 | SREF-0.3 | SREF | SREF + 0.3 | V |
| FEBAL1 | FBAL1 | Input such that pin $3=\operatorname{pin} 5$ and pin $4=\operatorname{pin} 6$, pin $18=$ SREF, Pin $76=5 \mathrm{~V}$, pin $19=$ SREF -1 V , pin 29: $\Delta$ GAIN | 3.7 | 6.7 | 9.7 | dB |
| FEBAL2 | FBAL2 | Input such that pin $3=\operatorname{pin} 5$ and pin $4=\operatorname{pin} 6$, pin $18=$ SREF, Pin $76=5 \mathrm{~V}$, pin $19=$ SREF +1 V , pin 29: $\Delta$ GAIN | -9.7 | -6.7 | -3.7 | dB |
| TEGAIN1 | TEG1 | Input to pin 10 and 11 , pin $18=$ SREF -0.75 V , pin $63=0 \mathrm{~V}$, Pin $68=0 \mathrm{~V}$, pin 36 | 11.8 | 15.8 | 19.8 | dB |
| TEGAIN2 | TEG2 | Input to pin 10 and 11 , pin $18=$ SREF +0.75 V , pin $63=0 \mathrm{~V}$, pin $68=0 \mathrm{~V}$, pin 36 | 26 | 29 | 32 | dB |
| TEOST | TOST | Pin $10=$ pin $11=$ PREF, pin $18=$ SREF, pin $63=0 \mathrm{~V}$ <br> Pin $68=0 \mathrm{~V}$, pin $53=5 \mathrm{~V}$, pin 36 | SREF-0.3 | SREF | SREF + 0.3 | V |
| TEBAL1 | TBAL1 | Input to pins 10 and 11 , pin $18=$ SREF, pin $63=0 \mathrm{~V}$ Pin $68=0 \mathrm{~V}$, pin $20=$ REF -1 V , pin $36 \Delta \mathrm{GAIN}$ | 4 | 7 | 10 | dB |
| TEBAL2 | TBAL2 | Input to pins 10 and 11, pin $18=$ SREF, pin $63=0 \mathrm{~V}$ Pin $68=0 \mathrm{~V}$, pin $20=$ REF +1 V , pin $36 \Delta$ GAIN | -10 | -7 | -4 | dB |
| DPD phase difference Voltage difference 1 | PD1 | The difference in the pin 30 voltage between when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $0^{\circ}$ phase, pin $3=5 \mathrm{MHz}$ with $36^{\circ}$ phase, and when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $36^{\circ}$ phase, pin $3=5 \mathrm{MHz}$ with $0^{\circ}$ phase. $\mathrm{R}_{\mathrm{L}}=6.8 \mathrm{k} \Omega$ | 0.25 | 0.37 | 0.48 | V |
| DPD phase difference Voltage difference 2 | PD2 | The difference in the pin 30 voltage between when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $0^{\circ}$ phase, pin $4=5 \mathrm{MHz}$ with $36^{\circ}$ phase, and when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $36^{\circ}$ phase, pin $4=5 \mathrm{MHz}$ with $0^{\circ}$ phase. $\mathrm{R}_{\mathrm{L}}=6.8 \mathrm{k} \Omega$ | -0.48 | -0.37 | -0.25 | V |

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| Parameter | Symbol | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | min | typ | max |  |
| DPD phase difference Voltage difference 3 | PD3 | The difference in the pin 30 voltage between when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $0^{\circ}$ phase, pin $5=5 \mathrm{MHz}$ with $36^{\circ}$ phase, and input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $36^{\circ}$ phase, pin $5=5 \mathrm{MHz}$ with $0^{\circ}$ phase. $\mathrm{R}_{\mathrm{L}}=6.8 \mathrm{k} \Omega$ | 0.25 | 0.37 | 0.48 | V |
| DPD phase difference Voltage difference 4 | PD4 | The difference in the pin 30 voltage between when input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $0^{\circ}$ phase, pin $6=5 \mathrm{MHz}$ with $36^{\circ}$ phase, and input with pin $1=$ SREF, pin $2=5 \mathrm{MHz}$ with $36^{\circ}$ phase, pin $6=5 \mathrm{MHz}$ with $0^{\circ}$ phase. $\mathrm{R}_{\mathrm{L}}=6.8 \mathrm{k} \Omega$ | -0.48 | -0.37 | -0.25 | V |
| DPD offset | DPDOF | Pin $1=$ SREF, $2=3=4=5=5 \mathrm{MHz}, \mathrm{R}_{\mathrm{L}}=6.8 \mathrm{k} \Omega$ | SREF - 0.5 | SREF | SREF + 0.5 | V |
| APC1 reference voltage | LDS1 | Pin $72=5 \mathrm{~V}$ | 150 | 180 | 200 | mV |
| APC1 off level | LDD1 | Pin $72=0 \mathrm{~V}$ | 4.5 | 5 |  | V |
| APC2 reference voltage | LDS2 | Pin $70=5 \mathrm{~V}$ | 150 | 180 | 200 | mV |
| APC2 off level | LDD2 | Pin $70=0 \mathrm{~V}$ | 4.5 | 5 |  | V |
| DFFTMU | DEFTMU | Pin $2=80 \mathrm{kHz}$. The time difference from a pin 2 edge to the rising edge on pin 46. | 1.8 | 4 | 5.8 | $\mu \mathrm{S}$ |
| DFFTMD | DFFTMD | Pin $2=80 \mathrm{kHz}$. The time difference from a pin 2 edge to the falling edge on pin 46. |  | 100 | 500 | $\mu \mathrm{S}$ |
| BPF1 | BPF1 | Pin $34=210 \mathrm{kHz}$, pin 32 | -1 | 2 | 5 | dB |
| BPF2 | BPF2 | Pin $34=120 \mathrm{kHz}$, pin 32 | -20 | -4 | 0 | dB |
| BPF3 | BPF3 | Pin $34=3500 \mathrm{kHz}$, pin 32 | -20 | -6.5 | -1 | dB |

## Functional Description

## 1. RF amplifier

- DVD mode

The RF signal input as a differential signal to pins 1 and 2 is passed through the RF VCA and is output from pin 65. The signal output from pin 65 is passed through a DVD RF equalizer and is output to later stage ICs from pins 40 and 41. When pin 63 is high, the RF signal does not pass through the customer amplifier. As a result, it is not influenced by the external peripheral circuit connected to pins 78 through 80. The RF VCA gain is controlled by applying a DC voltage to pin 21 . This IC provides two RF equalizer systems, and when pin 61 is high the pin 77 equalizer output is selected, and when low, the pin 52 equalizer output is selected. The amount of boost provided by the RF equalizer can be modified with the pin 22 DC voltage.

- CD mode

Pins 8,9 , and 10 can be set to be the RF input pins by setting pin 76 low. The customer amplifier connected to pins 78 through 80 is enabled by setting pin 63 low, a CD equalizer circuit can be constructed on pins 78 through 80 , and the signal will not pass through the DVD RF equalizer. The RF VCA gain is controlled by applying a DC voltage to pin 21.

## 2. Peak hold/bottom hold

The envelope waveforms for the peak and bottom of the front end RF signal output from pins 40 and 41 are output from pins 42 and 43. The envelope detection constants are set by the values of the resistors inserted between pins 43 and 39 and ground. The bottom hold detection constant can be increased by about a factor of 4 by setting pin 59 low. The bottom hold band width is also be increased by about a factor of 2 by setting pin 57 low.

## 3. Defect detection

The RF signal input from the pickup is converted to binary by a limiter circuit. The binary signal time is observed with a monostable multivibrator, and if there is no change in binary signal for over a certain fixed period, pin 46 is set high. The time period of the monostable multivibrator is set by the value of the capacitor connected to pin 47 and the resistor connected to pin 48.

## 4. RF equalizer

The CD RF equalizer is constructed from external components and the customer amplifier on pins 78 and 80, and outputs to the RF VCA in the next stage.
This IC provides two DVD RF equalizer systems, one of which is formed from external components and pins 67, 69, $71,73,75$, and 77 , and the other system is formed from external components and pins $52,54,56,58,60$, and 62 . When pin 61 is high, the equalizer system on pins $67,69,71,73,75$, and 77 is selected, and when that pin is low, the equalizer system on pins $52,54,56,58,60$, and 62 is selected. Since the customer amplifier is excluded from the signal path when pin 63 is high, the CD equalizer does not influence IC operation in DVD mode.

## 5. BCA

Peak envelope detection is applied to the previous stage RF signal output from pins 40 and 41 . The result is converted to binary by comparison with the BCA threshold and output from pin 45. The BCA threshold is input from external circuits to pin 25.

## 6. Reflect amplifier

The signals input to pins $3,4,5$, and 6 or pins 7,8 , and 9 are added with an summing amplifier. The pit component is removed from the input signal with a low-pass filter. The summed signal is passed through a VCA that adjusts the servo gain and output from pin 28. The VCA that adjusts the servo gain is controlled by the DC voltage applied to pin 18.

Note that when the pin 76 input is high, pins $3,4,5$, and 6 are selected, and when low, pins 7,8 , and 9 are selected.

## 7. FE amplifier

The signal input from either pins $3,4,5$, and 6 or pins 8 and 9 is first passed through an offset adjustment circuit and is then passed through a balance adjustment VCA. Then either the calculation < $\operatorname{pin} 3+\operatorname{pin} 5)-(\operatorname{pin} 4+\operatorname{pin} 6)>$ or <pin $8-$ pin $9>$ is performed. The result is passed through the servo gain adjustment VCA and output from pin 29. The gain of the balance adjustment VCA is adjusted by the DC voltage input to pin 19. The offset adjustment can be adjusted by the DC voltage applied to pin 23. The VCA that adjusts the servo gain is controlled by the DC voltage applied to pin 18.

Note that when the pin 76 input is high, pins 3, 4, 5, and 6 are selected, and when low, pins 8 and 9 are selected.
8. TE amplifier (for three-beam operation)

The current signal input to pin 10 and 11 is converted from a current to a voltage, passed through an offset adjustment circuit, and passed through a balance adjustment VCA. Then the calculation <pin 11 - pin 10> is performed and the result passed through a servo gain adjustment VCA. The result is output from pin 30 after band switching. The offset adjustment can be adjusted by the DC voltage applied to pin 24. The gain of the balance adjustment VCA can be adjusted by the DC voltage applied to pin 20. The VCA that adjusts the servo gain is controlled by the DC voltage applied to pin 18. The band switching circuit is a low-pass filter with a cutoff frequency of 30 kHz when pin 57 is high and 100 kHz when pin 57 is low. When pin 53 is low, pin 30 operates in hold mode.
Note that the three-beam TE is used when pin 68 is low.

## 9. DPD circuit

The phases of the signals input to pins 1 and 2 , and the signals input to pins $3,4,5$, and 6 are compared and the result outputs from pin 30.
The phase comparison result signal is output as a current by the pin 38 constant-current charge pump and converted to a voltage level by the external capacitor and resistor on pin 38. The voltage-converted signal is passed through a buffer amplifier, is band limited by a band switching circuit, and is output from pin 30. The charge pump is turned off when pin 51 is high. The band switching circuit is a low-pass filter with a cutoff frequency of 30 kHz when pin 57 is high and 100 kHz when pin 57 is low. When pin 53 is low, pin 30 operates in hold mode.
Note that the DPD circuit is used when pin 63 will be high.

## 10. PP amplifier

The signals input to pins $3,4,5$, and 6 are passed first through an offset adjustment circuit and then through a balance adjustment VCA. The calculation $\langle(\operatorname{pin} 3+\operatorname{pin} 6)-(\operatorname{pin} 4-\operatorname{pin} 5)>$ is performed. The result is passed through a servo gain adjustment VCA and is output from pin 35 after band switching. The offset adjustment can be adjusted by the DC voltage applied to pin 24 . The gain of the balance adjustment VCA can be adjusted by the DC voltage applied to pin 20. The VCA that adjusts the servo gain is controlled by the DC voltage applied to pin 18.

Note that the PP amplifier is used when pin 68 will be high.

## 11. Wobble bandpass filter

The signal input to pin 34 is passed through a bandpass filter and output from pin 32 . The frequency fo for the bandpass filter can be modified with the external resistor on pin 33 . When the pin 33 external resistor is $62 \mathrm{k} \Omega$, fo will be about 200 kHz .

## 12. APC circuit

A servo loop that holds the laser power fixed can be formed by inputting a monitor signal to pin 14 and connecting the laser driver to pin 13. The laser can be turned off by setting pin 72 low.
Note that there are two APC systems, with the other system consisting of pin 16 as the monitor input pin, pin 15 as the drive pin, and pin 70 as the laser off control pin.

## 13. Reference circuit

A voltage that is created by resistor dividing $\mathrm{V}_{\mathrm{CC}}$ by 2 is output from pin 26 . The pin 26 voltage is buffered and output from pins 79 and 27. The pin 79 voltage is a special-purpose reference voltage only for use by the pickup, and the pin 27 voltage is a reference supplied to the DSP and other systems.

## LA9702W

Pin Functions

| Pin No. | Pin | Function |
| :---: | :---: | :---: |
| 1 | RFN | RF signal - input |
| 2 | RFP | RF signal + input |
| 3 | PD1 | Pickup signal input |
| 4 | PD2 | Pickup signal input |
| 5 | PD3 | Pickup signal input |
| 6 | PD4 | Pickup signal input |
| 7 | PD5 | Pickup signal input |
| 8 | PD6 | Pickup signal input |
| 9 | PD7 | Pickup signal input |
| 10 | PD8 | Pickup signal input |
| 11 | PD9 | Pickup signal input |
| 12 | VCC | Power supply (servo signal system) |
| 13 | LDD1 | APC 1 output |
| 14 | LDS1 | APC 1 monitor voltage input |
| 15 | LDD2 | APC 2 output |
| 16 | LDS2 | APC 2 monitor voltage input |
| 17 | GND | Ground (servo signal system) |
| 18 | SGC | Servo gain control (RREC, FE, and TE) |
| 19 | FEBL | Focus balance adjustment |
| 20 | TEBL | Tracking balance adjustment |
| 21 | VGA | RF gain adjustment |
| 22 | BST | Equalizer boost adjustment |
| 23 | FOST | Focus offset adjustment |
| 24 | TOST | Tracking offset adjustment |
| 25 | BCATH | BCA threshold adjustment |
| 26 | REFI | Reference voltage setting |
| 27 | SREF | Servo signal reference voltage output |
| 28 | RREC | Reflection output |
| 29 | FE | Focus error output |
| 30 | TE | Tracking error output |
| 31 | THC | TE hold time constant setting capacitor connection |
| 32 | WO | Wobble output |
| 33 | ISET | Bandpass filter center frequency setting resistor connection |
| 34 | WOI | Push-pull signal input |
| 35 | WOO | Push-pull signal output |
| 36 | TEO | Three-beam TE gain setting |
| 37 | TEN | Three-beam TE gain setting |
| 38 | CP | Charge pump gain setting resistor and capacitor connection |
| 39 | BHI | Bottom hold detection constant setting resistor connection |
| 40 | RFON | RF - output |
| 41 | RFOP | RF + output |
| 42 | BH | RF bottom detection output |
| 43 | PHI | Peak hold detection constant setting resistor connection |
| 44 | PH | RF peak detection output |
| 45 | BCA | BCA output |
| 46 | DEF | Defect output (High: defect detected) |
| 47 | DEFC | Defect detection capacitor connection |
| 48 | TC | Defect detection constant setting resistor connection |
| 49 | GND | Ground (DPD system) |
| 50 | LPC | RF DC servo capacitor connection |
| 51 | CPOF | Charge pump on/off control (High: off) |
| 52 | EQO2 | RF equalizer setting |
| 53 | TH | Tracking hold (High: hold) |

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Continued from preceding page.

| Pin No. | Pin |  |
| :---: | :---: | :--- |
| 54 | EQL2 | RF equalizer setting |
| 55 | VCC | Power supply (DPD system) |
| 56 | EQO4 | RF equalizer setting |
| 57 | XHTR | Tracking and bottom detection band switching (Low: high band) |
| 58 | EQ14 | RF equalizer setting |
| 59 | XQBH | Bottom detection time constant switching (Low: fast) |
| 60 | EQ03 | RF equalizer setting |
| 61 | EQSCT | Equalizer switching (High: pin 77 selected, low: pin 52 selected) |
| 62 | EQI3 | RF equalizer setting |
| 63 | DPD/TE | DPD/three-beam tracking switching (High: DPD) |
| 64 | RFO2 | RF output |
| 65 | RFO1 | RF output |
| 66 | VCC | Power supply (RF system) |
| 67 | EQ11 | RF equalizer setting |
| 68 | PP/TE | Three-beam/push-pull tracking switching (Low: three-beam) |
| 69 | EQO1 | RF equalizer setting |
| 70 | LDON2 | APC 2 laser on/off control (High: on) |
| 71 | EQl2 | RF equalizer setting |
| 72 | LDON1 | APC 1 laser on/off control (High: on) |
| 73 | EQO2 | RF equalizer setting |
| 74 | GND | Ground (RF system) |
| 75 | EQL1 | RF equalizer setting |
| 76 | RFSCT | RF input switching (High: RF differential input, PP error) |
| 77 | EQO1 | RF equalizer setting |
| 78 | CAO | Customer amplifier output |
| 79 | PREF | Reference voltage output (pickup) |
| 80 | CAN | Customer amplifier input |

Note: The equalizer constants support $1 \times$ and $2 \times$ speeds.



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